

# Digital Communications 1

## Written exam on October 17, 2019

Institute of Telecommunications

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**Please note:**

- You may use the official lecture notes, a pocket calculator, and a collection of mathematical formulas.
- Personal notes, materials from exercise classes, and pre-calculated problems may not be used.
- Legible writing and a clear layout of your derivations and solutions are absolutely necessary!
- Provide detailed derivations. When using results from the lecture notes, they must be explicitly referenced.

### Problem 1 (20 credits)

Consider three base stations  $B_i$ ,  $i = 1, 2, 3$ , transmitting an identical symbol sequence  $a[k]$  to a single receiver. The base stations use passband PAM transmission with symbol alphabet  $\mathcal{A} = \{-1, 1\}$  and equally likely transmit symbols. The Fourier transform of the transmit pulses is given by  $G_i(j\omega) = \sqrt{G_i R(j\omega)}$ , where  $G_1 = 0.002$ ,  $G_2 = 0.01$ ,  $G_3 = 0.005$  and  $R(j\omega)$  is the Fourier transform of a raised-cosine pulse. The three channels between the base stations and the receiver are AWGN channels with identical power spectral density  $S_N(j\omega) = 0.5 \cdot 10^{-3} \text{W/Hz}$ , no inter-channel interference, and statistically independent noise processes. The receiver uses a bank of three sampled matched filters (i.e., one filter for each base station signal) whose outputs are the pseudosymbol sequences  $q_i[k]$ ,  $i = 1, 2, 3$ , followed by a detector.

- a) Specify and sketch the equivalent discrete-time baseband PAM system with the single input  $a[k]$  and the three outputs  $q_i[k]$ ,  $i = 1, 2, 3$ . In particular, calculate the total pulses  $p_i[k]$  for channels  $i = 1, 2, 3$ . Are the individual transmissions ISI-free?

We will now consider two different detector designs.

- b) *Selection combining*: The detector uses only a single pseudosymbol sequence  $q_i[k]$  to decide upon the transmitted symbol sequence. Determine the ML decision rules and calculate the symbol error probabilities for the three choices  $i = 1, 2, 3$ . Which choice results in the smallest symbol error probability?
- c) *Optimum combining*: The detector processes all three pseudosymbol sequences to make a decision about the transmitted symbol sequence. Derive the ML detector for this case. (*Hint*: use the joint channel pdf  $f(q_1[k], q_2[k], q_3[k] | a[k])$ .) Sketch an implementation that uses only one symbol-wise slicer.
- d) Calculate the symbol error probability of the optimal receiver from c) and compare it with the symbol error probability of the receiver from b). Interpret the results.

## Problem 2 (20 credits)

Consider replacing the zero-forcing equalizer

$$D_{\text{ZF}}(e^{j\theta}) = \frac{1}{1 - 2 \cos(\theta)}$$

by a decision feedback equalizer. The transmit symbols of the system are taken from the alphabet  $\{1, -1\}$  with equal probabilities. The symbol sequence and the equivalent discrete-time noise  $Z[k]$  are uncorrelated and both white. The noise is zero-mean, and the noise variance is  $\sigma_Z^2 = 1$ .

- Calculate the pulse  $p[k]$  of the equivalent discrete-time baseband channel consisting of transmit pulse, physical channel, and matched receive filter.
- Assume a general feedforward filter  $d[k], k \in [-L, L]$ . Calculate the overall pulse  $p^{(d)}[k]$  at the output of the feedforward filter. How long is this pulse?
- Consider now the case  $L = 1$ . What is the minimum length  $K$  of the feedback filter such that as much ISI as possible is canceled (under the assumption that all previous symbol decisions were correct)?
- Calculate the coefficients of the MSE-optimum feedforward filter  $\mathbf{d}_{\text{MSE}}$ .

*Hint: The inverse of a matrix*

$$\mathbf{A} = \begin{pmatrix} a_{11} & a_{12} & a_{13} \\ a_{21} & a_{22} & a_{23} \\ a_{31} & a_{32} & a_{33} \end{pmatrix}$$

*can be calculated as*

$$\mathbf{A}^{-1} = \frac{1}{\det \mathbf{A}} \begin{pmatrix} a_{22}a_{33} - a_{23}a_{32} & a_{13}a_{32} - a_{12}a_{33} & a_{12}a_{23} - a_{13}a_{22} \\ a_{23}a_{31} - a_{21}a_{33} & a_{11}a_{33} - a_{13}a_{31} & a_{13}a_{21} - a_{11}a_{23} \\ a_{21}a_{32} - a_{22}a_{31} & a_{12}a_{31} - a_{11}a_{32} & a_{11}a_{22} - a_{12}a_{21} \end{pmatrix},$$

*with*

$$\det \mathbf{A} = a_{11}a_{22}a_{33} + a_{12}a_{23}a_{31} + a_{13}a_{21}a_{32} - a_{13}a_{22}a_{31} - a_{12}a_{21}a_{33} - a_{11}a_{23}a_{32}.$$

**Problem 3 (20 credits)**

Consider transmission of a symbol  $A \in \{-1/2, 1\}$ . At the slicer input, the receiver observes a random variable  $Q$  that is corrupted by real-valued additive noise  $Z$ :

$$Q = A + Z.$$

The noise is exponentially distributed, i.e.,

$$f_Z(z) = \begin{cases} e^{-z} & \text{if } z \geq 0, \\ 0 & \text{if } z < 0. \end{cases}$$

- a) For each  $a \in \{-1/2, 1\}$ , calculate and sketch  $f_{Q|A}(q|a)$ .
- b) Calculate the probabilities  $P\{Q > \eta_1 | A = -1/2\}$  and  $P\{Q < \eta_2 | A = 1\}$  for  $\eta_1 > -1/2$  and  $\eta_2 < 1$ .
- c) Consider ML detection. Determine the decision rule and sketch the decision intervals. Calculate the conditional symbol error probability  $P\{\mathcal{E} | A = a\}$  for each  $a \in \{-1/2, 1\}$ .
- d) Consider a detector using the following decision rule:

$$\hat{A} = \begin{cases} -1/2 & \text{if } -1/2 \leq Q \leq 0, \\ 1 & \text{if } 1 \leq Q \leq 2, \\ \text{undecided} & \text{otherwise.} \end{cases}$$

Sketch the decision intervals. Calculate the conditional symbol error probabilities  $P\{\hat{A} = 1 | A = -1/2\}$  and  $P\{\hat{A} = -1/2 | A = 1\}$ .

**Problem 4 (20 credits)**

Consider a variant of FSK using the transmit pulses

$$g_m(t) = G \cos((2m + 1)\omega_0 t) \operatorname{sinc}(\omega_0 t), \quad m = 1, \dots, M_a$$

where  $\operatorname{sinc}(x) = \frac{\sin(x)}{x}$  and  $\omega_0 = \frac{\pi}{T_s}$ .

- a) Calculate and sketch the Fourier transform  $G_m(j\omega)$ .
- b) Show that all shifted transmit pulses  $g_m(t - kT_s)$  and  $g_n(t - lT_s)$  ( $k, l \in \mathbb{Z}$ ) are orthogonal unless both  $k = l$  and  $n = m$ .
- c) Calculate and sketch the transmit bandwidth  $B_T$  and the spectral efficiency  $\nu$  as a function of  $M_a$ .