

# Digital Communications 1

## Written exam on September 22, 2020

Institute of Telecommunications

TU Wien

**Please note:**

- You may use the official lecture notes, a pocket calculator, and a collection of mathematical formulas.
- Personal notes, materials from exercise classes, and pre-calculated problems may not be used.
- Legible writing and a clear layout of your derivations and solutions are absolutely necessary!
- Provide detailed derivations. When using results from the lecture notes, they must be explicitly referenced.

**Problem 1 (20 credits)**

Consider passband PAM transmission using the signal constellation

$$\mathcal{A} = \{-1 \pm j, 1 \pm j, \beta \pm j\},$$

with  $\beta > 1$ . The receiver performs symbolwise ML detection. The symbols  $a[k] \in \mathcal{A}$  are corrupted by additive noise  $z[k] \in \mathbb{C}$ , yielding the following signal at the slicer input:

$$q[k] = a[k] + z[k],$$

with real part  $q_{\text{R}}[k]$  and imaginary part  $q_{\text{I}}[k]$ . The real part and the imaginary part of the noise at the slicer input,  $z[k] = z_{\text{R}}[k] + jz_{\text{I}}[k]$ , are statistically independent zero-mean Gaussian random variables with variances  $\sigma_{\text{R}}^2 = 1$  and  $\sigma_{\text{I}}^2 = 0.5$ , respectively.

- Sketch the signal constellation and the decision regions of the symbolwise ML detector.
- Assume that the symbol  $1 - j$  was transmitted. Specify the ML decision region for a correct decision by specifying the intervals within which  $q_{\text{R}}[k]$  and  $q_{\text{I}}[k]$  must lie.
- Calculate the probability of a correct decision for the case that  $1 + j$  was transmitted, depending on  $\beta$ . Find the value of  $\beta$  for which this probability is 0.6.
- The prior probabilities of symbols with a positive real part is twice as large as that of symbols with a negative real part. Using the value of  $\beta$  calculated in part c), calculate the mean symbol power  $P_A$ .

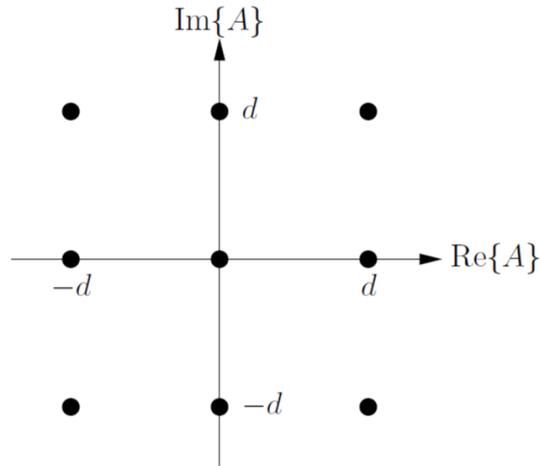
**Problem 2 (20 credits)**

Consider a passband PAM system with symbol period  $T_s$ , transmit pulse  $g(t)$  with energy  $E_g = 1$ , and transmit symbols  $A[k]$ . The receive pulse in the equivalent baseband is  $h(t) = g(t) + \alpha g(t - 2T_s)$  with  $|\alpha| < 1$ . The additive channel noise has a power spectral density of  $N_0/2$ . The receiver consists of a sampled matched filter, a noise whitening filter (based on a *causal* factorization), a slicer, and an equalizer.

- a) Determine the impulse response, transfer function, and noise power spectral density of the equivalent discrete-time model of the transmission system including the noise whitening filter but not the equalizer.
- b) Calculate the zero-forcing equalizer (placed after the noise whitening filter). Compare the noise variance at the input and the output of the equalizer.
- c) Verify that the output of the noise whitening filter is free of precursor intersymbol interference.
- d) Determine the decision-feedback equalizer that completely removes postcursor intersymbol interference. (The noise whitening filter is used as the feedforward filter of the decision-feedback equalizer.)

**Problem 3 (20 credits)**

Consider the transmission of a single symbol  $A$  from the following signal constellation:



The transmit probabilities are

$$P\{A = \pm d\} = P\{A = \pm jd\} = 1/10 \quad \text{and} \quad P\{A = (-1 \pm j)d\} = P\{A = (1 \pm j)d\} = 1/20.$$

Note that the transmit probability  $P\{A = 0\}$  was not specified and must be deduced. The received symbol is  $Q = A + Z$ , where  $Z$  is circularly symmetric complex Gaussian noise with variance  $\sigma_Z^2 = d^2/10$ . Symbol  $A$  and noise  $Z$  are statistically independent.

- a) Calculate the symbol mean  $\mu_A$  and the symbol variance  $\sigma_A^2$ .
- b) Calculate and sketch the decision regions of the MAP receiver.
- c) Calculate the error probability of the MAP receiver.
- d) Calculate the error probability of the ML receiver.

**Problem 4 (20 credits)**

Consider a variant of FSK using the transmit pulses

$$g_m(t) = G \cos((2m + 1)\omega_0 t) \operatorname{sinc}(\omega_0 t), \quad m = 1, \dots, M_a$$

where  $\operatorname{sinc}(x) = \frac{\sin(x)}{x}$  and  $\omega_0 = \frac{\pi}{T_s}$ .

- a) Calculate and sketch the Fourier transform  $G_m(j\omega)$ .
- b) Show that all shifted transmit pulses  $g_m(t - kT_s)$  and  $g_n(t - lT_s)$  ( $k, l \in \mathbb{Z}$ ) are orthogonal unless both  $k = l$  and  $n = m$ .
- c) Calculate and sketch the transmit bandwidth  $B_T$  and the spectral efficiency  $\nu$  as a function of  $M_a$ .