

Modulations- und Detektionsverfahren

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Please note:

- You may use the official lecture notes, a pocket calculator, and a collection of mathematical formulas.
- Personal notes, materials from exercise classes, and pre-calculated problems may not be used.
- Legible writing and a clear layout of your derivations and solutions are absolutely necessary!
- Provide detailed derivations. When using results from the lecture notes, they must be explicitly referenced.

Problem 1 (20 credits)

Consider the following pulse set:

$$\begin{aligned}g_1(t) &= \operatorname{sinc}\left(\pi \frac{t}{T_s}\right) \cos\left(3\pi \frac{t}{T_s}\right) \\g_2(t) &= \operatorname{sinc}\left(\frac{\pi}{2} \frac{t}{T_s}\right) \cos\left(\frac{3\pi}{2} \frac{t}{T_s}\right) \\g_3(t) &= \operatorname{sinc}\left(\frac{\pi}{2} \frac{t}{T_s}\right) \cos\left(3\pi \frac{t}{T_s}\right).\end{aligned}$$

- a) Which of the pulses satisfy the Nyquist criterion? (A derivation is required.)
- b) Show that those pulses $g_i(t)$ that satisfy the Nyquist criterion also satisfy the generalized Nyquist criterion.
- c) Using the same pulses as in b), consider orthogonal multipulse PAM with symbol alphabet $\mathcal{A} = \{-1, 1\}$. The symbols $A_m[k]$ are uniformly distributed and uncorrelated, i.e., $\mathbb{E}\{A_m[k]A_n^*[l]\} = \delta_{k,l}\delta_{m,n}$. Sketch the power spectral density $S_{\bar{S}}(j\omega)$ of the stationarized transmit signal $\bar{S}(t)$.

Problem 2 (20 credits)

Consider transmission of the signals

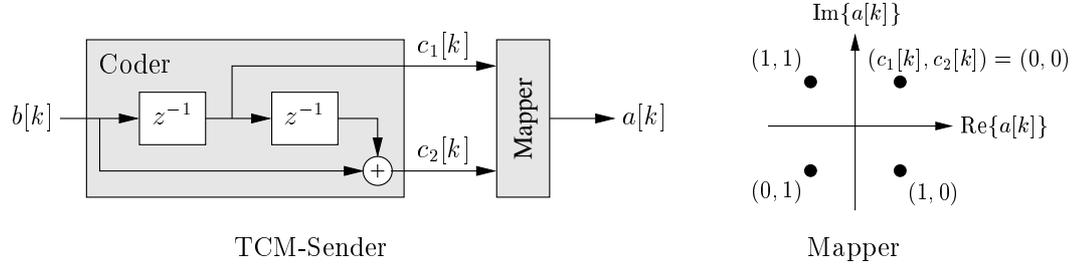
$$s^{(i)}(t) = is(t), \quad i \in \{1, 2, 3\} \quad \text{with } p_I(1) = p_I(2) = 1/4$$

over an AWGN channel with noise variance $N_0/2$. Here, $s(t)$ is a fixed signal with energy $E_s \neq 0$.

- a) Determine an orthonormal basis of the transmit signal space.
- b) Specify the vectors corresponding to the signals $s^{(i)}(t)$.
- c) Sketch an implementation of the ML detector using inner products of the received signal with the basis functions.
- d) Sketch an implementation of the ML detector using inner products of the received signal with the transmit signals.
- e) Calculate the conditional sequence error probabilities and the unconditional sequence error probability of the ML detector.

Problem 3 (20 credits)

Consider a system with trellis coded modulation (TCM). The transmitter consists of a convolutional encoder and a symbol mapper, as illustrated below.



The convolutional encoder uses the data bits $b[k]$ to produce two code bits at each step,

$$c_1[k] = b[k-1] \quad \text{and} \quad c_2[k] = b[k] \oplus b[k-2]$$

(\oplus denotes addition modulo 2). The symbol mapper maps these code bits to symbols $a[k]$ (with $|a[k]|^2 = 2$).

The received signal is given by

$$y[k] = a[k] + w[k]$$

(ISI-free channel), with white complex Gaussian noise $w[k]$.

- a) For the bit sequence $\{1, 1, 0, 1, 0\}$, derive the corresponding code bits $c_1[k]$, $c_2[k]$ and transmit symbols $a[k]$ (assuming that all previous data bits $b[k]$ are zero).
- b) Make a list of all possible combinations of $b[k-2]$, $b[k-1]$, and $b[k]$ and the resulting $a[k]$.
- c) Sketch a trellis diagram of the TCM transmitter.
- d) The transmitted bit sequence is detected by an ML receiver that is implemented using the Viterbi algorithm. Calculate the detected bits $\hat{b}[k]$ from the received signal $y[0] = 3.1 - j1.4$, $y[1] = 0.2 + j1.2$, $y[2] = -0.7 + j0.9$ (assuming that $b[k] = 0$ for $k < 0$).

Problem 4 (20 credits)

Consider a PAM system with received pulse

$$h(t) = e^{-at} \mathbf{u}(t) + e^{-a(t-T_S)} \mathbf{u}(t - T_S), \quad \text{where } a > 0.$$

Here, T_S is the symbol rate and $\mathbf{u}(t)$ denotes the unit step function. The noise $N(t)$ is white with power spectral density $N_0/2$. The receiver consists of a bandpass-lowpass transformation, a receive filter $f(t)$, a symbol-rate sampler, a linear ZF equalizer, and a slicer.

- a) Specify the receive filter $f(t)$ that minimizes the noise enhancement of the ZF equalizer. Use this receive filter in the following.
- b) Find a function $\phi(t)$ such that $h(t) = \phi(t) + \phi(t - T_S)$. Use this relation for the following calculations.
- c) Calculate the equivalent discrete-time pulse $p[k]$ at the equalizer input.
- d) Find the value of a such that $p[1] = 2p[2]$.