

# Digital Communications 1

## Written exam on May 3, 2016

Institute of Telecommunications  
Vienna University of Technology

**Please note:**

- You may use the official lecture notes, a pocket calculator, and a collection of mathematical formulas.
- Personal notes, materials from exercise classes, and pre-calculated problems may not be used.
- Legible writing and a clear layout of your derivations and solutions are absolutely necessary!
- Provide detailed derivations. When using results from the lecture notes, they must be explicitly referenced.

### Problem 1 (20 credits)

A stationary and white bit sequence (bit rate  $R_b = 64$  kbit/s) is to be transmitted over a bandlimited AWGN channel (power spectral density  $S_N(j\omega) = N_0/2 = 9 \cdot 10^{-7}$  W/Hz, bandwidth  $B_T = 24$  kHz). The modulation scheme is passband PAM using an  $M_a$ -ary QAM constellation. The Fourier transform of the transmit pulse is given by  $G(j\omega) = \sqrt{R(j\omega)}$ , where  $R(j\omega)$  is the Fourier transform of a raised-cosine pulse with roll-off factor  $\alpha = 0.5$ .

- a) Determine the size of the symbol alphabet  $M_a$  and sketch the corresponding symbol constellation.
- b) Design a receive filter for ISI-free transmission.
- c) Sketch the block diagram of the ML receiver.

For the following, assume that symbols with positive real part are twice as likely than symbols with negative real part.

- d) Calculate the minimum symbol distance  $d_a$  for a symbol error probability of  $P\{\mathcal{E}_s\} = 10^{-7}$ . Use for this the approximation  $P\{\mathcal{E}_s\} \approx \bar{\mathcal{N}}\mathcal{Q}\left(\frac{d_{\min}}{\sqrt{2}\sigma_Z}\right)$ .
- e) How does the symbol constellation need to be shifted to minimize the mean transmit power  $P_{\bar{S}}$ ? Sketch the shifted constellation. How large is the achieved reduction of  $P_{\bar{S}}$ ?

## Problem 2 (20 credits)

Consider equalization of a channel with the equivalent discrete-time baseband pulse

$$p[k] = \delta[k] + \delta[k - 1] + \delta[k + 1].$$

The transmit symbols are taken from the alphabet  $\{1, -1\}$  with equal probabilities. The symbol sequence and the equivalent discrete-time noise  $Z[k]$  are uncorrelated and both white. The noise is zero-mean, and the noise variance is  $\sigma_Z^2 = 1$ .

- a) Calculate  $D_{\text{ZF}}(e^{j\theta})$ .
- b) Now, assume that an MSE equalizer  $d_{\text{MSE}}[k], k \in [-L, L]$  is used. What is a suitable value for  $L$ ? Why?
- c) Sketch the block diagram of an MSE equalizer using  $L$  as determined in b).
- d) Calculate  $d_{\text{MSE}}[k]$  for  $k \in [-L, L]$ .

*Hint:* The inverse of a matrix

$$\mathbf{A} = \begin{pmatrix} a_{11} & a_{12} & a_{13} \\ a_{21} & a_{22} & a_{23} \\ a_{31} & a_{32} & a_{33} \end{pmatrix}$$

can be calculated as

$$\mathbf{A}^{-1} = \frac{1}{\det \mathbf{A}} \begin{pmatrix} a_{22}a_{33} - a_{23}a_{32} & a_{13}a_{32} - a_{12}a_{33} & a_{12}a_{23} - a_{13}a_{22} \\ a_{23}a_{31} - a_{21}a_{33} & a_{11}a_{33} - a_{13}a_{31} & a_{13}a_{21} - a_{11}a_{23} \\ a_{21}a_{32} - a_{22}a_{31} & a_{12}a_{31} - a_{11}a_{32} & a_{11}a_{22} - a_{12}a_{21} \end{pmatrix},$$

with

$$\det \mathbf{A} = a_{11}a_{22}a_{33} + a_{12}a_{23}a_{31} + a_{13}a_{21}a_{32} - a_{13}a_{22}a_{31} - a_{12}a_{21}a_{33} - a_{11}a_{23}a_{32}.$$

**Problem 3 (20 credits)**

A BPSK symbol  $S \in \{-1, 1\}$  with prior probability  $p_S(1) = p_1$  is corrupted by additive noise  $N \in \mathbb{R}$  that is statistically independent of  $S$  and distributed as follows:

$$f_N(n) = \begin{cases} \alpha(2+n) & \text{if } -2 \leq n \leq -1, \\ \alpha & \text{if } -1 \leq n \leq 0, \\ \alpha(1-n/2) & \text{if } 0 \leq n \leq 2, \\ 0 & \text{otherwise.} \end{cases}$$

Thus, the received random variable is  $Y = S + N$ .

- a) Sketch  $f_N(n)$  and calculate  $\alpha$ .
- b) Sketch  $f_{Y|S}(y|s)$  for  $s \in \{-1, 1\}$ . Calculate the ML decision threshold  $\eta_{\text{ML}}$ .
- c) Calculate the conditional error probabilities  $P_{\text{ML}}\{\mathcal{E}|S = s\}$  for  $s \in \{-1, 1\}$  and the unconditional error probability  $P_{\text{ML}}\{\mathcal{E}\}$  of the ML detector.
- d) Calculate the MAP decision threshold  $\eta_{\text{MAP}}$  as a function of  $p_1$ .
- e) Calculate the value of  $p_1$  for which
  - e1)  $\eta_{\text{MAP}} = \eta_{\text{ML}}$ ;
  - e2)  $\eta_{\text{MAP}} = 0$ ;
  - e3)  $\eta_{\text{MAP}} = -2/3$ .

**Problem 4 (20 credits)**

A sequence of symbols  $a[k] \in \{-1, 0, 1\}$ , with  $a[k] = 0$  for  $k < 0$ , is transmitted over a discrete-time channel with impulse response  $h[k] = \delta[k] - 0.5\delta[k-1]$ . The additive noise is white and Gaussian. The received sequence  $y[k]$  is given by  $y[0] = 0.8$ ,  $y[1] = 0.2$ , and  $y[2] = -1.1$ .

- a) Represent this channel by a shift register circuit, a state diagram, and an elementary stage of the corresponding trellis diagram.
- b) Use the Viterbi algorithm for ML sequence detection. Which sequence  $\hat{a}[k]$  ( $k = 0, 1, 2$ ) is obtained with this receiver?
- c) An alternative receiver uses a zero-forcing equalizer followed by a slicer. Which sequence  $\hat{a}[k]$  ( $k = 0, 1, 2$ ) is obtained with this receiver? You may assume that  $y[k] = 0$  for  $k < 0$ .