

Guidelines

- Write your name and matriculation number on each sheet of paper.
- Only clearly readable exercise-elaborations are evaluated.
- Results have to be provided together with an evident way of calculation.
- Keep textual answers short and concise. Lengthy or vague statements won't gain points.

Exercise 1.1 (0.5 points)

The bilinear transform is a widely used method for designing digital filters from an already designed analog filter. It can be interpreted as a mathematical transformation from the s -domain (laplace transform) to the z -domain (z -transform):

$$H(z) = H(s) \Big|_{s=\frac{2}{T} \frac{z-1}{z+1}} .$$

We first investigate on how the transformation maps certain regions of the s -plane onto the z -plane.

1. Calculate r and ϕ of the term $z = re^{j\phi}$ in terms of the real part σ and the imaginary part ω of $s = \sigma + j\omega$.
2. Consider the general value $s = \sigma + j\omega$. Determine the corresponding mappings from the s -plane to the z -plane for $\sigma > 0$, $\sigma = 0$ and $\sigma < 0$, respectively (which regions of the s -plane are mapped onto which regions of the z -plane?). Display your findings by appropriate sketches.
3. For $\sigma = 0$, s becomes $s = j\omega$. Determine the relation between the frequencies ω in the analog system and the frequencies Ω in the digital system.
4. Given the transfer function of an analog filter

$$H(s) = \frac{s+2}{(s+1)(s+3)}, \quad (1)$$

design the corresponding digital filter, by applying the bilinear transform. Determine

- (a) the corresponding transfer function $H(z)$,
- (b) the discrete-time impulse response $h[n]$.

Hint: Employ a partial fraction expansion.

Exercise 1.2 (0.5 points)

1. Show that each system with transfer function

$$H(e^{j\Omega}) = A(e^{j\Omega})e^{-j\alpha\Omega+j\beta} \quad (2)$$

and real-valued $A(e^{j\Omega})$ has linear phase and constant group delay. Explicitly calculate phase and group delay of such a system.

2. In literature we distinguish between four different types of FIR filters with linear phase, whose impulse responses h_n are defined by:

- (a) $h_n = h_{M-n} \quad 0 \leq n \leq M \quad M \text{ even}$
- (b) $h_n = h_{M-n} \quad 0 \leq n \leq M \quad M \text{ odd}$
- (c) $h_n = -h_{M-n} \quad 0 \leq n \leq M \quad M \text{ even}$
- (d) $h_n = -h_{M-n} \quad 0 \leq n \leq M \quad M \text{ odd.}$

For each of the four systems, sketch an exemplary impulse response (For this task choose $M = 5$ and $M = 6$, respectively).

3. Now assume four general h_n , which satisfy the properties of the four filter types defined above. Calculate the four resulting transfer functions for an arbitrary M (M can be assumed to be even or odd, according to the used filter type). Further, determine phase and group delay for each filter type.

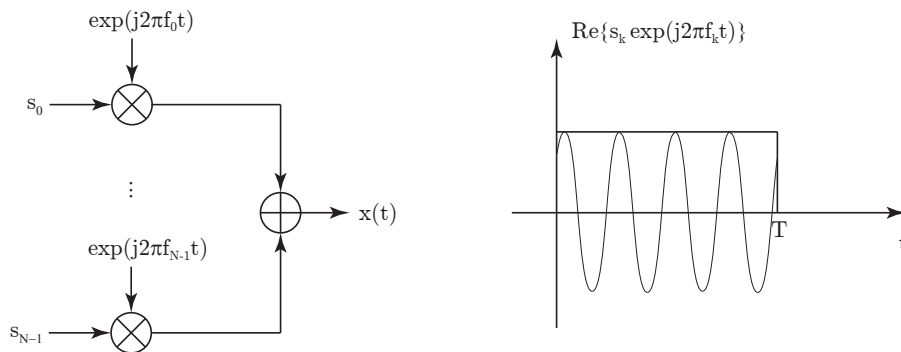


Figure 1.1: Signal modulation

Exercise 1.3 (0.5 points)

A stream of complex valued symbols is split into blocks of N symbols each. The N symbols of each block are modulated on N carriers f_k with constant frequency spacing Δf . The signal is windowed with a window length T , as indicated in Figure 1.1.

1. Let s_k denote the k -th symbol of a block, with $k = 0, \dots, N - 1$. Provide a low-pass equivalent description of the signal $x(t)$, i.e., set $f_0 = 0$. Hint: Express f_k in terms of Δf .
2. Let $\Delta f = \frac{1}{NT_s}$ and assume that the signal is sampled at $t = nT_s$, where T_s denotes the sampling period. Provide a discrete time description $x[n]$ of the signal.
3. Given T_s , how do you have to choose T such that the sampled signal is Intersymbol Interference (ISI) free within a block?
Hint: Consider the Nyquist ISI criterion.
4. Which type of linear transformation is performed here on the symbols of each block?
5. Reformulate the discrete time description of $x[n]$ into vector-matrix notation, assuming $\underline{s} = [s_0, \dots, s_{N-1}]^T$ and $\underline{x} = [x_0, \dots, x_{N-1}]^T$ as the input- and output-vectors, respectively. Then, answer the following questions:
 - (a) Are the columns of the resulting matrix linearly independent?
 - (b) Are the columns of the resulting matrix mutually orthogonal?
Hint: Consider the geometric series

$$\sum_{n=0}^{N-1} e^{j2\pi(k-k')n/N} \quad (3)$$

- (c) *Normalize the matrix such that each vector (i.e., column) in the matrix has length 1.*

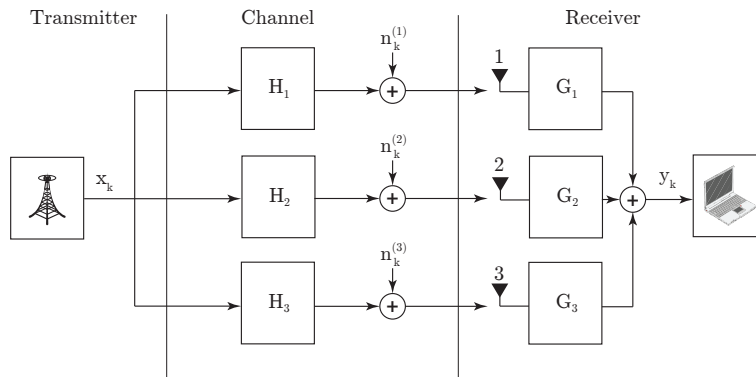


Figure 1.2

MATLAB-Exercise 1.1 (1 points)

Consider a SIMO (Single Input Multiple Output) experimental setup, as shown in Figure 1.2. The scenario is assumed to be static and the channels are known exactly from prior measurements:

$$H_1(q^{-1}) = 1 - 0.25q^{-2} \quad (4)$$

$$H_2(q^{-1}) = 1 - q^{-1} + 0.25q^{-2} \quad (5)$$

$$H_3(q^{-1}) = 1 + 2q^{-1} \quad (6)$$

The target of the experiment is to receive a distortion-free signal $y_k = x_{k-1}$ (where x_k denotes the input signal). This should be achieved with as few antennas as possible. The corresponding receive filters for each antenna are given as

$$G_i(q^{-1}) = g_0^{(i)} + g_1^{(i)}q^{-1} \quad (7)$$

1. Preliminaries

In the first part of the exercise, the noise is assumed as $n_k^{(i)} = 0, \forall i$.

(a) Use MATLAB to determine the filter coefficients $g_k^{(i)}$ for the cases:

- i. Only antenna 1 is used.
- ii. Antenna 1 and 2 are used.
- iii. All three antennas are used.

If a calculation is not possible, reason your answer in detail. Note:

- You can use the "\ " operator in MATLAB (see documentation for further details)
- Verify your results since MATLAB will generate output for all three cases.

(b) Can one receive antenna be turned off without changing the output behavior of the system? Reason your answer.

2. Simulations

In the second part of the exercise, the scenario as shown in Figure 1.2 is simulated. The target is to investigate the behavior of the system under various SNR (Signal-to-Noise-Ratio) conditions with the receive filters as obtained in the first part.

Preparation:

- Generate a stream of random symbols x_k from the alphabet $\{1+j, -1+j, -1-j, 1-j\}$, i.e., each symbol codes for 2 bits. All symbols are equally likely and $|x_k|^2 = 1$.
- Generate a stream of properly normed noise samples for each channel. The samples are assumed to be white complex Gaussian.

(a) For an SNR of 10 dB

- i. Provide a scatterplot of the input signal of receive filter 3.
- ii. Draw a scatterplot of the receive signal after filtering (y_k in the figure).

(b) For an SNR-range -10 dB \dots 20 dB

- i. Decode the stream of received symbols by Maximum-Likelihood (ML) detection, i.e., the detected symbol is obtained as $\text{sign}(\Re(y_k)) + j \text{sign}(\Im(y_k))$, where $\text{sign}(\cdot)$ denotes the sign function

$$\text{sign}(x) = \begin{cases} -1, & x < 0 \\ 0, & x = 0 \\ 1, & x > 0. \end{cases} \quad (8)$$

- ii. Provide a plot of the Bit Error Ratio (BER) versus Signal to Noise Ratio (SNR) in double-logarithmic scale (In order to get meaningful values for the BER, choose the number of simulated symbols sufficiently large).